

Description

The AP61041 is a low-current, synchronous buck converter providing high efficiency, excellent transient response, and high DC output accuracy. The ultra-low I_Q of the AP61041 makes it ideal for applications such as fitness wearables, health monitoring, Bluetooth, and other handheld devices. The AP61041 is optimized to operate with a 2.2μH inductor and 10μF input and output capacitors. The device provides a 400mA output current with an input voltage range of 2.15V to 5.5V.

The constant on-time control scheme handles wide input/output voltage ratios and provides a low external component count with outstanding performance in line and load transient responses. It also has seamless transitions between buck and 100% duty cycle modes.

The AP61041 provides two VSEL pins for programmable feedback voltages from 0.6V to 1.8V. The device also features undervoltage lockout (UVLO), overcurrent protection (OCP), and overtemperature protection (OTP) to protect the circuit.

This device is available in a small 1.17mm x 0.77mm, 6-ball, WLCSP package.

Features

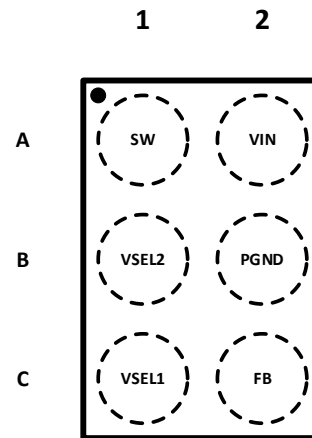
- VIN 2.15V to 5.5V
- Preselected Output Voltages for Reduced Component Count
- 400mA Continuous Output Current, Efficiency Up to 95%
- 1.7μA (2.5μA Max) I_Q to Maximize Light-Load Efficiency
- 1.1MHz Switching Frequency
- Pulse-Frequency Modulation
- Fully Protected from Undervoltage Lockout, Overcurrent, Short Circuit, and Overtemperature Conditions
- **Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)**
- **Halogen and Antimony Free. "Green" Device (Note 3)**
- **For automotive applications requiring specific change control (i.e. parts qualified to AEC-Q100/101/104/200, PPAP capable, and manufactured in IATF 16949 certified facilities), please [contact us](https://www.diodes.com/quality/product-definitions/) or your local Diodes representative.**

<https://www.diodes.com/quality/product-definitions/>

- Notes:
1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.
 2. See <https://www.diodes.com/quality/lead-free/> for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.

Pin Assignments

TOP VIEW



X1-WLB1208-6

Applications

- Wearables
- Fitness trackers
- Smart watches
- Health monitors
- Bluetooth low-energy devices
- Ultra-low power applications
- Energy harvesting

Typical Application Circuit

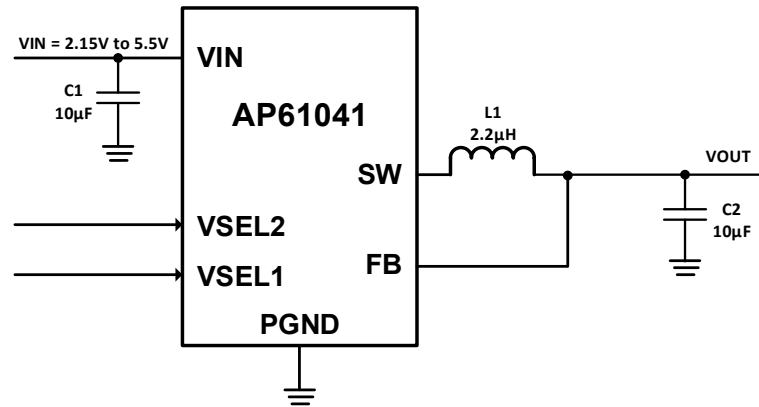


Figure 1. Typical Application Circuit

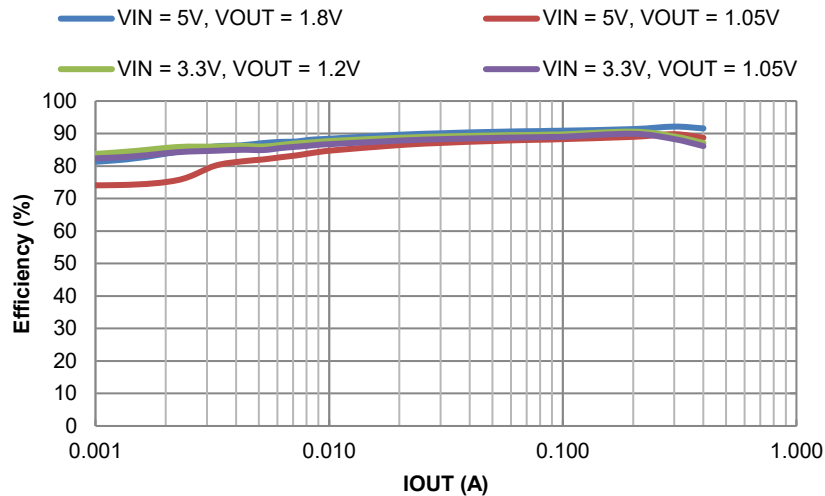


Figure 2. PFM Efficiency vs. Output Current

Pin Descriptions

Pin Name	Pin Number 6 BALLS	Function
SW	A1	Switch node.
VIN	A2	Power input and supply for logic control circuitries.
VSEL2	B1	Output voltage select pin 2. VSEL1 is a three-state input used to select the output voltage. It can be dynamically changed during operation. See the Setting Output Voltage section for more details.
PGND	B2	Power ground.
VSEL1	C1	Output voltage select pin 1. VSEL2 is a three-state input used to select the output voltage. It can be dynamically changed during operation. See the Setting Output Voltage section for more details.
FB	C2	Feedback sensing terminal for the output voltage.

Functional Block Diagram

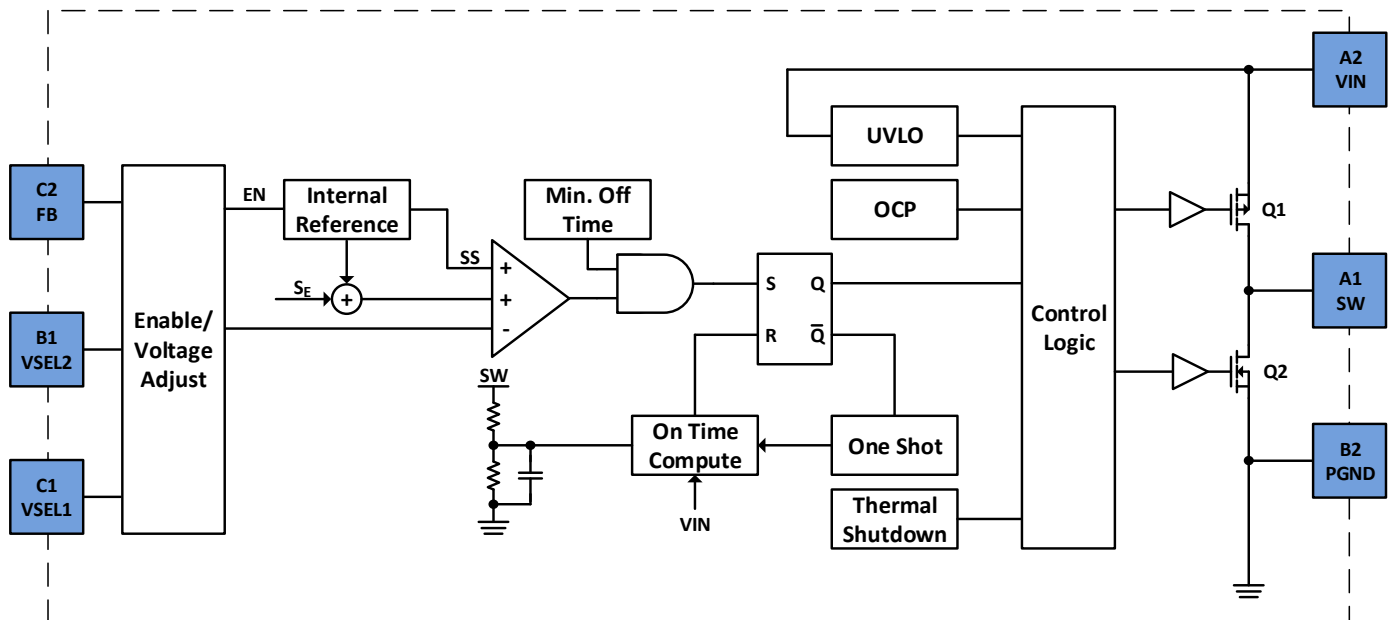


Figure 3. Functional Block Diagram

Absolute Maximum Ratings (Note 4) (@T_A = +25°C, unless otherwise specified.)

Symbol	Parameter	Rating	Unit
V _{VIN}	Supply Voltage	-0.3 to +7.0	V
V _{SW}	Switch Node Voltage	-1.0 to V _{VIN} + 0.3 (DC)	V
V _{SW}	Switch Node Voltage	-2.5 to 8 (20ns)	V
All other pins		-0.3 to +7.0	V
T _J	Junction Temperature	+150	°C
T _L	Lead Temperature	+260	°C
ESD Susceptibility (Note 5)			
HBM	Human Body Mode	3000	V
CDM	Charged Device Model	1000	V

- Notes:
- Stresses greater than the 'Absolute Maximum Ratings' specified above may cause permanent damage to the device. These are stress ratings only; functional operation of the device at these or any other conditions exceeding those indicated in this specification is not implied. Device reliability may be affected by exposure to absolute maximum rating conditions for extended periods of time.
 - Semiconductor devices are ESD sensitive and may be damaged by exposure to ESD events. Suitable ESD precautions should be taken when handling and transporting these devices.

Thermal Resistance

Symbol	Parameter	Package	Rating, JEDEC (Note 6)	AP61041 EVM (Note 7)	Unit
θ _{JA}	Junction to Ambient	WLB1208-6	100	60	°C/W
θ _{JC(TOP)}	Junction to Case (Top)	WLB1208-6	49	35	°C/W
θ _{JB}	Junction to Board (bottom)	WLB1208-6	15.7	12	°C/W
ψ _{JT}	Junction to Top Characterization Parameter	WLB1208-6	4.6	1.2	°C/W
ψ _{JB}	Junction to Board Characterization Parameter	WLB1208-6	15	12	°C/W
θ _{JC(BOT)}	Junction to Case (Bottom)	WLB1208-6	-	-	°C/W

- Notes:
- Device mounted on FR-4 substrate, JEDEC 4 layer 50mm x 50mm PCB board (2 oz copper), with minimum recommended pad layout.
 - Device mounted on Diodes evaluation board. See user guide for more detail.

Recommended Operating Conditions (Note 8) (@T_A = +25°C, unless otherwise specified.)

Symbol	Parameter	Min	Max	Unit
V _{IN}	Supply Voltage	2.15	5.5	V
I _{OUT}	Output Current, 2.15V<V _{IN} <2.5V, V _{OUT} <1.8V	-	300	mA
	Output Current, V _{IN} >2.5V	-	400	
T _J	Operating Junction Temperature	-40	+125	°C

- Note:
- The device function is not guaranteed outside of the recommended operating conditions.

Electrical Characteristics (At $T_J = +25^\circ\text{C}$, $V_{IN} = 3.6\text{V}$, $V_{EN} = 3.6\text{V}$ unless otherwise specified. Min/Max limits apply across the recommended operating junction temperature range, -40°C to $+125^\circ\text{C}$, and input voltage range, 2.15V to 5.5V, unless otherwise specified.)

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
I_{SHDN}	Shutdown Supply Current	$V_{IN} = 5.5\text{V}$, $V_{VSEL2} = \text{OPEN}$, $V_{VSEL1} = \text{OPEN}$	-	440	800	nA
I_Q	Supply Current (Quiescent)	$V_{IN} = V_{VSEL2} = V_{VSEL1} = 5.5\text{V}$ $V_{FB} = 1\text{V}$	-	1.2	2.0	μA
		$I_{OUT} = 0\text{A}$	-	1.1	1.6	μA
POR/UVLO	V_{VIN} Power On Reset Voltage Threshold, Rising Edge	-	-	2.07	2.15	V
	V_{VIN} , Undervoltage Lockout Threshold, Falling Edge	-	-	1.90	2.00	V
	Hysteresis	-	-	170		mV
$R_{DS(ON)1}$	High-Side Switch On-Resistance from VIN to SW	-	-	321	440	m Ω
$R_{DS(ON)2}$	Low-Side Switch On-Resistance from SW to PGND	-	-	150	210	m Ω
$R_{Discharge}$	VOUT Soft Discharge On-Resistance	-	-	50	-	Ω
I_{LIMIT}	Positive HS Current Limit, Q1	-	600	800	-	mA
I_{LIMIT_VALLEY}	LS Current Limit, Q2 Current from Source to Drain	-	-	470	-	mA
I_{NLIMIT}	Negative LS Current Limit, Q2 Current from Drain to Source	-	-	400	-	mA
F_{SW}	Oscillator Frequency	-	-	1.1	-	MHz
V_{FB}	Feedback Voltage	$V_{VSEL2} = 0$, $V_{VSEL1} = 0$	1.035	1.050	1.065	V
		$V_{VSEL2} = 0$, $V_{VSEL1} = 1$	0.888	0.9	0.912	V
		$V_{VSEL2} = 1$, $V_{VSEL1} = 0$	0.690	0.7	0.710	V
		$V_{VSEL2} = 1$, $V_{VSEL1} = 1$	0.615	0.625	0.635	V
		$V_{VSEL2} = 0$, $V_{VSEL1} = \text{OPEN}$	0.591	0.6	0.609	V
		$V_{VSEL2} = \text{OPEN}$, $V_{VSEL1} = 0$	1.180	1.2	1.220	V
		$V_{VSEL2} = 1$, $V_{VSEL1} = \text{OPEN}$	1.280	1.3	1.320	V
		$V_{VSEL2} = \text{OPEN}$, $V_{VSEL1} = 1$	1.780	1.8	1.820	V
V_{VSEL2}	VSEL2 Logic LOW	-	-	-	0.4	V
	VSEL2 Logic OPEN	-	0.5	-	0.9	V
	VSEL2 Logic HIGH	-	1.4	-		V
V_{VSEL1}	VSEL1 Logic LOW	-	-	-	0.4	V
	VSEL1 Logic OPEN	-	0.5	-	0.9	V
	VSEL1 Logic HIGH	-	1.4	-	-	V
T_{SS}	Soft-Start Period	-	-	0.8	-	ms
T_{SD}	Thermal Shutdown (Note 9)	-	-	150	-	$^\circ\text{C}$
T_{Hys}	Thermal Hysteresis (Note 9)	-	-	15	-	$^\circ\text{C}$

Note: 9. Compliance to the datasheet's limits is assured by one or more of these methods: production tests, characterization, and/or by design.

Typical Performance Characteristics (AP61041 @ TA = +25°C, VIN = 5V, VOUT = 1.8V, PFM, unless otherwise specified.)
(continued)

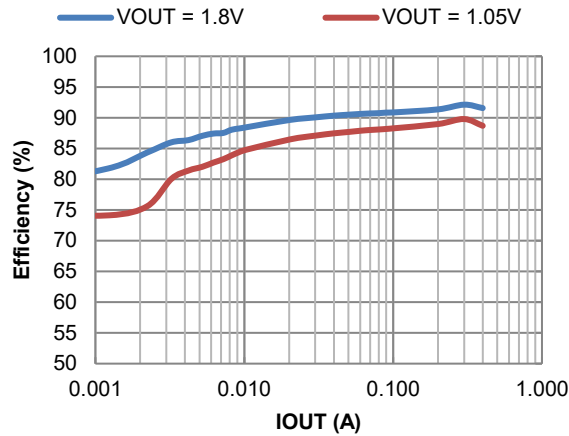


Figure 4. Efficiency vs. Output Current, VIN = 5V

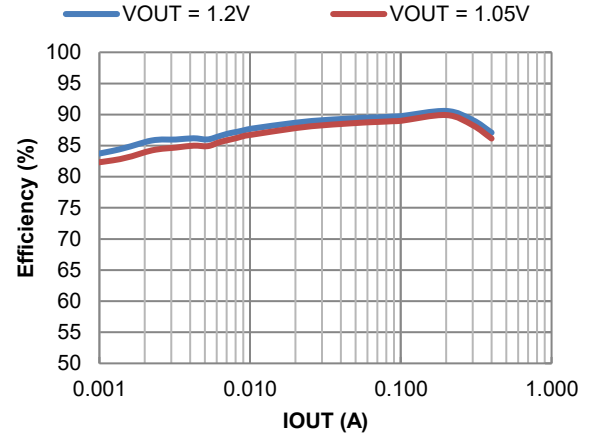


Figure 5. Efficiency vs. Output Current, VIN = 3.3V

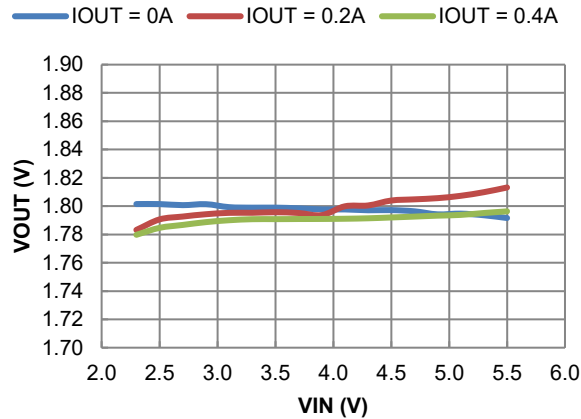


Figure 6. Line Regulation

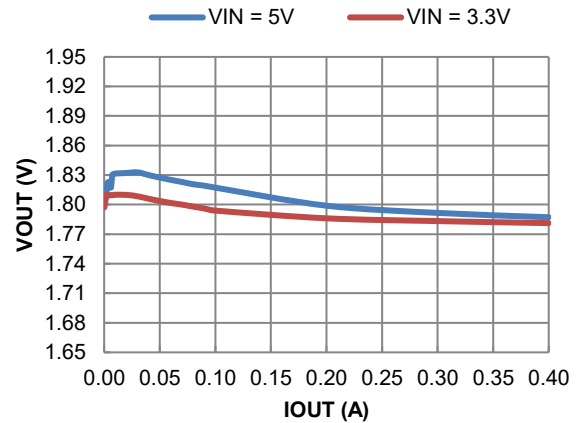


Figure 7. Load Regulation

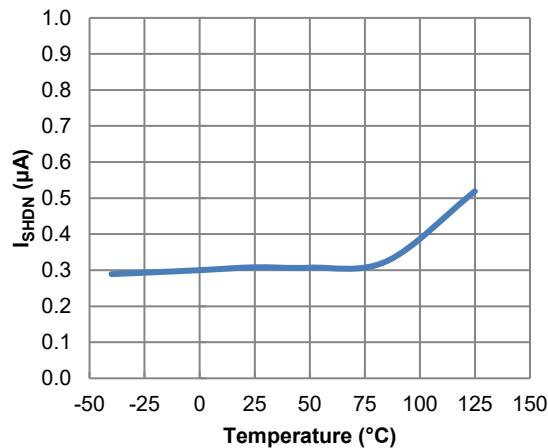


Figure 8. ISHDN vs. Temperature

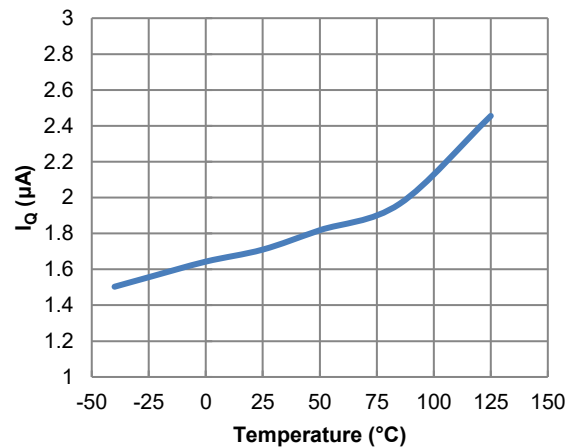


Figure 9. IQ vs. Temperature

Typical Performance Characteristics (AP61041 @ TA = +25°C, VIN = 5V, VOUT = 1.8V, PFM, unless otherwise specified.)
(continued)

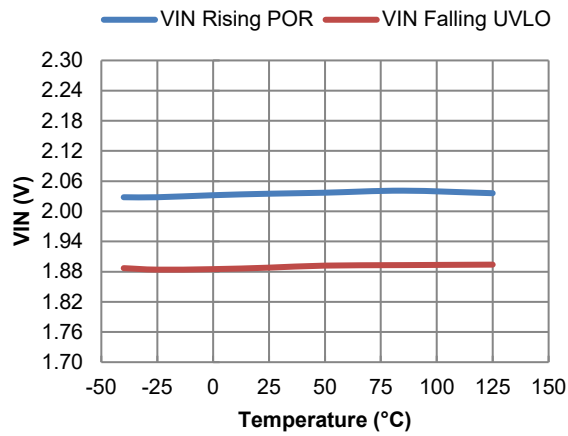


Figure 10. VIN Power-On Reset and UVLO vs. Temperature

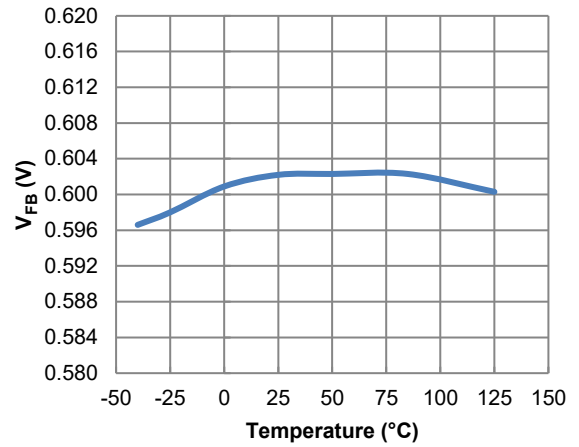


Figure 11. Feedback Voltage vs. Temperature

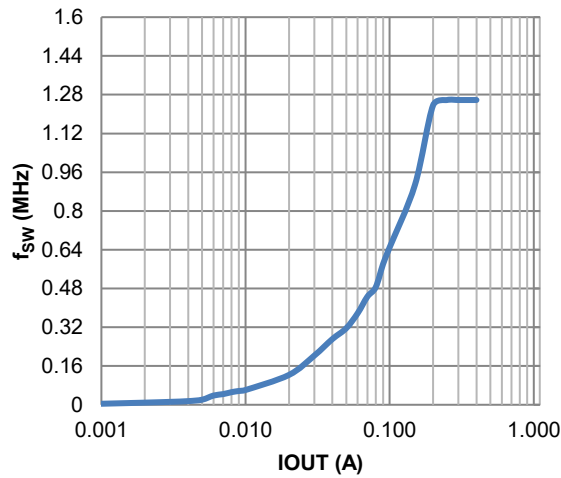


Figure 12. fsw vs. Load

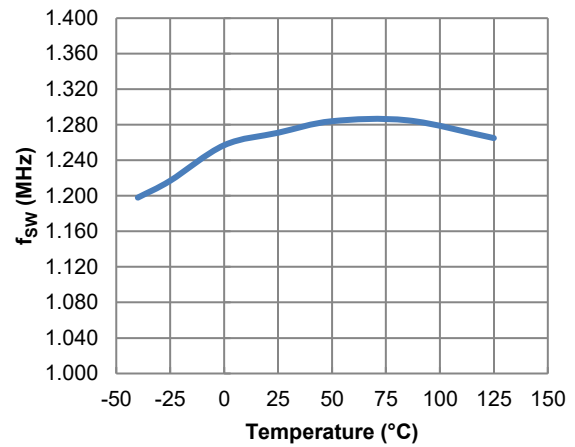


Figure 13. fsw vs. Temperature

Typical Performance Characteristics (AP61041 @ TA = +25°C, VIN = 5V, VOUT = 1.8V, PFM, unless otherwise specified.)
(continued)

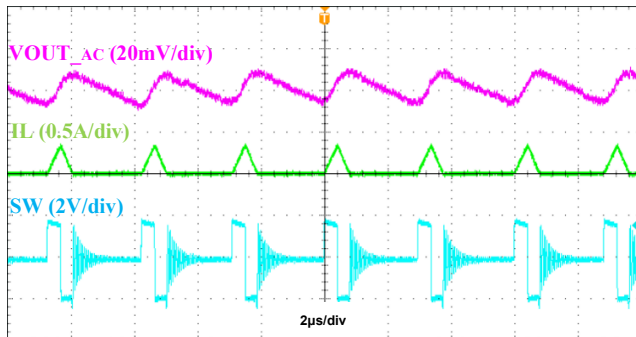


Figure 14. Output Voltage Ripple, IOUT = 50mA

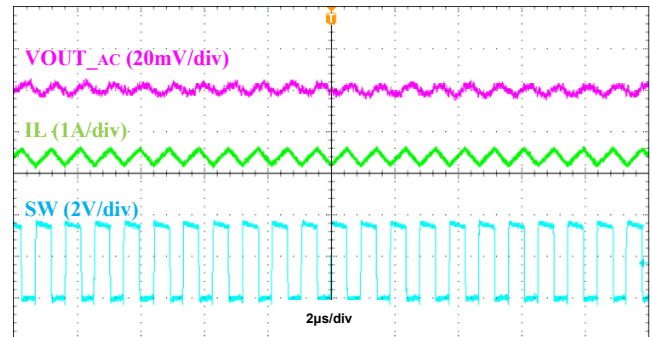


Figure 15. Output Voltage Ripple, IOUT = 0.4A

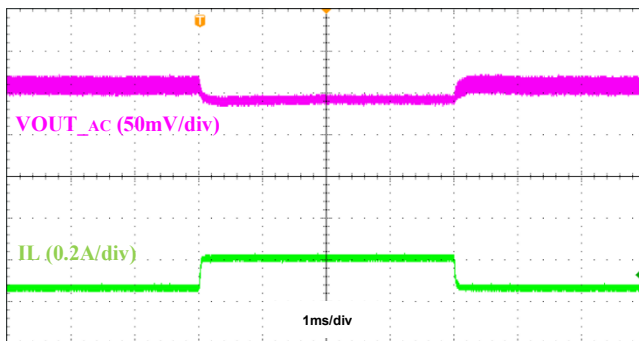


Figure 16. Load Transient, IOUT = 20mA to 200mA to 50mA

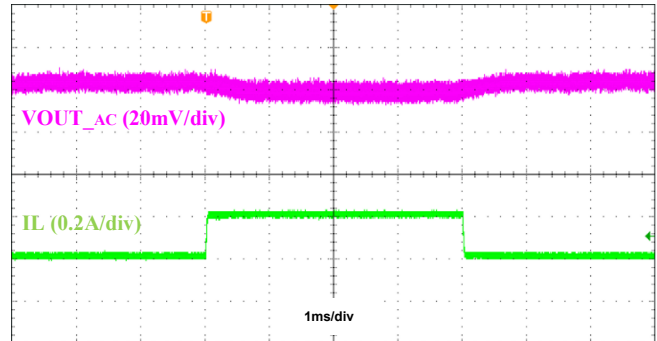


Figure 17. Load Transient, IOUT = 200mA to 400mA to 200mA

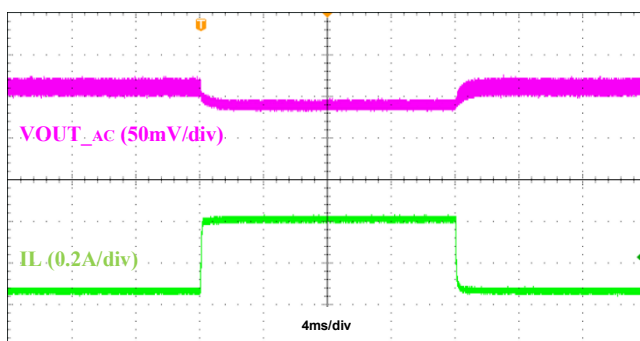


Figure 18. Load Transient, IOUT = 50mA to 400mA to 50mA

Typical Performance Characteristics (AP61041 @ $T_A = +25^\circ\text{C}$, $V_{IN} = 5\text{V}$, $V_{OUT} = 1.8\text{V}$, PFM, unless otherwise specified.)
(continued)

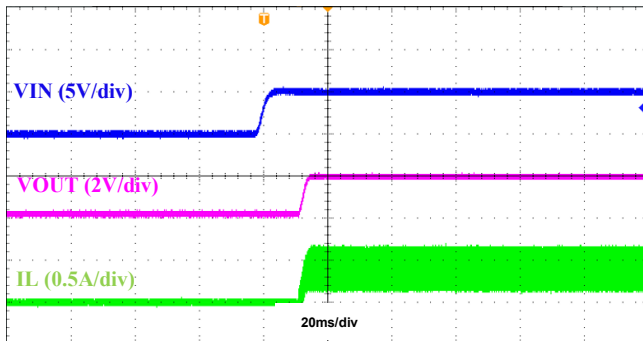


Figure 19. Startup Using VIN, IOU = 0.4A

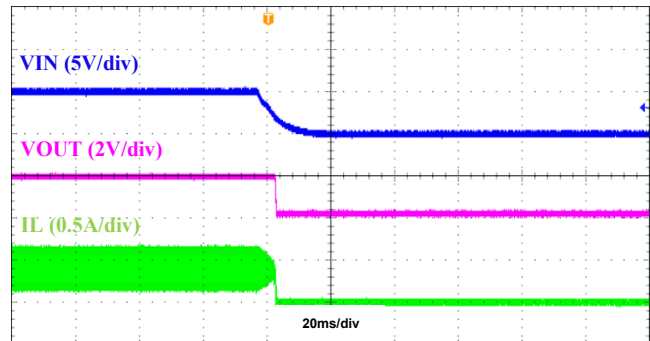


Figure 20. Shutdown Using EN, IOU = 0.4

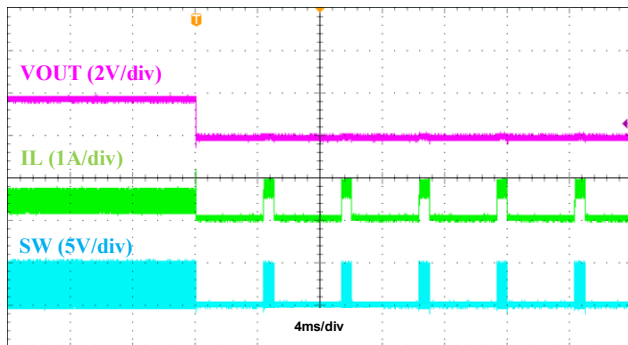


Figure 21. Output Short Protection, IOU = 0.4A

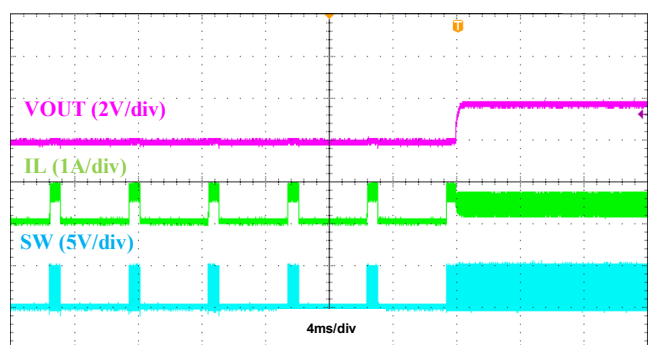


Figure 22. Output Short Protection Recovery, IOU = 0.4A

Application Information

Theory of Operation

The AP61041 is a 400mA, adaptive, constant on-time control, synchronous buck regulator with integrated power MOSFETs. Constant on-time control assures excellent line regulation, load regulation, and a wide apparent loop bandwidth for fast response to load transients. Figure 1 and figure 3 depict the typical application circuit and functional block diagram of the AP61041, respectively. The buck controller drives the internal high-side power PMOS, Q1, and low-side power NMOS, Q2. The buck regulator can operate from an unregulated DC source, such as a battery, with a voltage ranging from 2.15V to 5.5V. The converter output can be regulated as low as 0.6V, to as high as 5.5V. There are preselected feedback voltages that are programmed by the VSEL1 and VSEL2 pins. See the **Setting Output Voltage** section for more details.

Chip Enable Control

Chip enable is controlled by the combination of VSEL1 and VSEL2 pins. VSEL1 and VSEL2, when both being OPEN, will disable the regulator. Any other combination of VSEL1 and VSEL2 will enable the regulator.

Setting Output Voltage

The AP61041 has two voltage select pins, VSEL1 and VSEL2, each with three input logic states (LOW, OPEN, and HIGH). The logic combinations of VSEL1 and VSEL2 allow for eight preselected feedback voltage levels, with the ninth combination disabling the device. Table 1 summarizes the device actions given a particular VSEL1 and VSEL2 combination. Connect the FB pin directly to the output to have the V_{OUT} equal to the V_{FB} .

VSEL2	VSEL1	Device Enabled	V_{FB} (V)
LOW	LOW	Yes	1.05
LOW	HIGH	Yes	0.9
HIGH	LOW	Yes	0.7
HIGH	HIGH	Yes	0.625
LOW	OPEN	Yes	0.6
OPEN	LOW	Yes	1.2
HIGH	OPEN	Yes	1.3
OPEN	HIGH	Yes	1.8
OPEN	OPEN	No	-

Table 1. Preselected Feedback Voltage Combinations

The maximum transition time of the VSEL1 and VSEL2 must be less than 100 μ s to avoid an undefined voltage state. In addition to the pre-selected feedback voltages, the output voltage can be adjusted from the feedback voltage using an external resistive divider, as seen in figure 23. A CFF of 39pF must be added to maintain output stability.

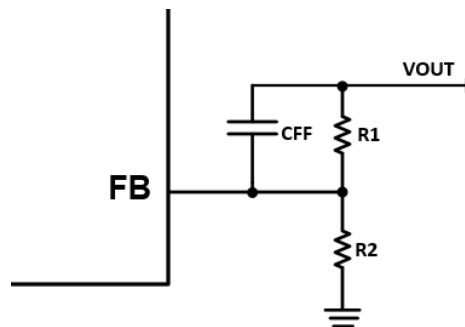


Figure 23. External Feedback Resistive Network

Given the selected feedback voltage (V_{FB}), the selected R2 value of approximately 100k Ω , and the target output voltage (V_{OUT}), the value of R1 can be determined by the following equation:

$$R1 = \frac{R2 * (V_{OUT} - V_{FB})}{V_{FB}} \quad \text{Eq.1}$$

Application Information

Overcurrent Protection

The AP61041 detects the current limit on the high-side power MOSFET, Q1, to protect the device against overload and short-circuit conditions. The peak current in the switch is monitored cycle-by-cycle with a comparator delay of approximately 100ns to guard against noise glitches. If the high-side Q1 current limit is reached, the high-side Q1 is turned off and the low-side Q2 is turned on until the switch current decreases below OC threshold. The frequency is reduced to protect the device from damage. The Q1 peak current limit remains active during this state. After 17 consecutive cycles in this OCP event, the regulator enters hiccup mode where all power FETs are turned off and will wait for 15ms before attempting to restart.

Reverse Current Protection

During fixed-frequency operation, a reverse-current comparator on switch Q2 monitors the current entering VOUT. When this current exceeds 400mA (typical), switch Q2 will be turned off for the remainder of the switching cycle. This feature protects the buck converter from excessive reverse current if the buck output is held above the regulation voltage by an external source.

Undervoltage Lockout

The undervoltage lockout (UVLO) feature prevents abnormal operation if the supply voltage is too low to guarantee proper operation. When the VIN voltage falls below the UVLO threshold, the regulator is disabled.

Thermal Shutdown

A built-in thermal protection feature protects the AP61041 if the die temperature reaches +150°C (typical). At this die temperature, the regulator is completely shut down. The die temperature continues to be monitored in this thermal-shutdown mode. When the die temperature falls to +125°C (typical), the device will resume normal operation. When exiting thermal shutdown, the AP61041 will execute its soft-start sequence.

Output Active Discharge

The buck provides an internal 50Ω resistor for output active discharge function. The internal resistor discharges the energy stored in the output capacitor to PGND whenever the regulator is disabled. When the regulator remains enabled, the internal resistor is disconnected from the output.

Inductor Selection

An inductor with a high-frequency core material (e.g., ferrite core) should be used to minimize core losses and provide good efficiency. The inductor must be able to handle the peak switching currents without saturating. A 2.2μH inductor with ≥600mA saturation current rating is recommended. Select an inductor with a low DCR to provide good efficiency. In applications where radiated noise must be minimized, a toroidal or shielded inductor can be used.

VIN and VOUT Capacitor Selection

The input and output capacitors should be a ceramic X5R type with low ESL and ESR. The recommended input capacitor value is 10μF, as this provides adequate RMS current to minimize the input voltage ripple. A minimum of 10μF is required to maintain full functionality of the part. The recommended output capacitor is 10μF, 10V, X5R. Note that the effective value of a ceramic capacitor derates with DC voltage bias across it. This decreasing may be up to 70% of the rated capacitance. Refer to the capacitor datasheet to ensure the combined effective output capacitance is at least 30μF for proper operation over the entire recommended load current range. Low output capacitance may lead to large output voltage drop during load transient or unstable operation.

Application Information

PC Board Layout

The AP61041 works at 400mA load current. A 2oz Copper in both the top and bottom layer is recommended. Correct PCB layout is critical for proper operation of the AP61041. The following are some general guidelines for the recommended layout:

1. The input and output capacitors should be positioned directly across VIN-PGND and as close to the IC as possible to ensure noise-free operation.
2. The ground connections of the input and output capacitors should be kept as short as possible. The objective is to minimize the current loop between the ground pads of the input and output capacitors and the PGND pins of the IC. Use via, if required, to take advantage of a PCB ground layer underneath the regulator.
3. Fill the second layer with PGND. Single point connects the GND to the second layer's PGND.
4. Minimize the trace lengths on the feedback loop to avoid switching noise pick-up. Via should be avoided on the feedback loop to minimize the effect of board parasitic, particularly during load transients.
5. The SW trace should be short.
6. See figure 24 below for more details on the recommended layout.

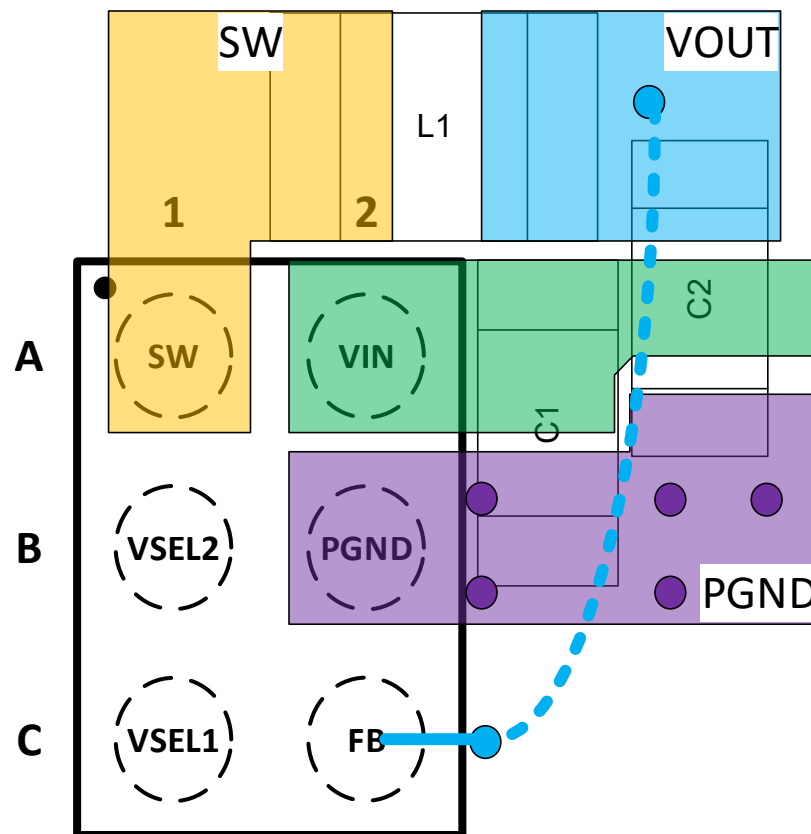
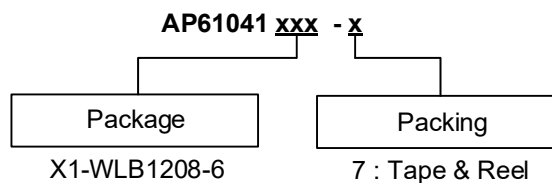


Figure 24. Recommended Layout Design

Ordering Information

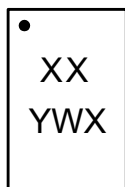


Orderable Part Number	Package	Package Code	Packing		
			Quantity	Carrier	Part Number Suffix
AP61041CP6-7	X1-WLB1208-6	CP6	3,000	Tape and Reel	-7

Marking Information

X1-WLB1208-6

(Top View)



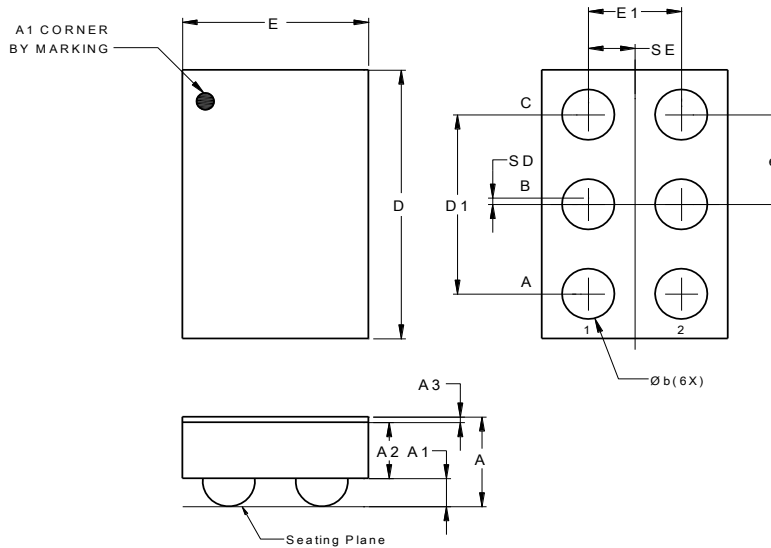
XX : Identification Code
 Y : Year : 0~9
 W : Week : A~Z : 1~26 week;
 a~z : 27~52 week; z represents
 52 and 53 week
 X : Internal Code

Orderable Part Number	Package	Identification Code
AP61041CP6-7	X1-WLB1208-6	KC

Package Outline Dimensions

Please see <http://www.diodes.com/package-outlines.html> for the latest version.

X1-WLB1208-6

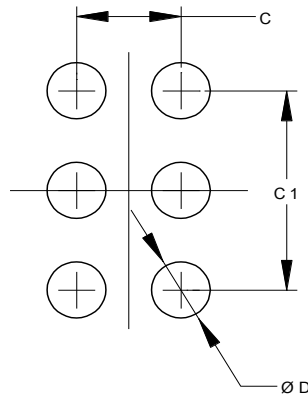


X1-WLB1208-6			
Dim	Min	Max	Typ
A	0.400	0.500	0.450
A1	0.155	0.195	0.175
A2	0.225	0.275	0.250
A3	0.020	0.030	0.025
b	0.200	0.260	0.230
D	1.160	1.220	1.190
D1	0.770	0.830	0.800
E	0.760	0.820	0.790
E1	0.370	0.430	0.400
e	0.400 BSC		
SD	0.000 BSC		
SE	0.200 BSC		
All Dimensions in mm			

Suggested Pad Layout

Please see <http://www.diodes.com/package-outlines.html> for the latest version.

X1-WLB1208-6



Dimensions	Value (in mm)
C	0.400
C1	0.800
D	0.250

Mechanical Data

- Moisture Sensitivity: Level 1 per J-STD-020
- Terminals: Finish – Matte Tin Plated Leads, Solderable per MIL-STD-202, Method 208e3
- Weight: 0.7 mg (Approximate)

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