



## PAM8013

### **3W MONO CLASS D AUDIO AMPLIFIER**

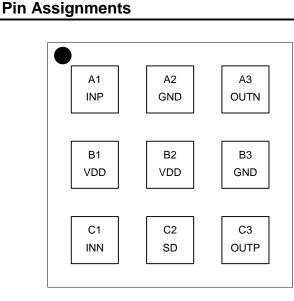
### Description

The DIODES<sup>™</sup> PAM8013 is a 3W mono, filter-less, Class-D amplifier with high PSRR and a differential input that eliminates noise and RF rectification.

Features like its 90% efficiency and small PCB area make the PAM8013 Class-D amplifier ideal for cellular handsets. The filter-less architecture simplifies application design and requires no external output filter, fewer external components, less PCB area, and lower system costs.

The PAM8013 features short-circuit protection and thermal shutdown.

The PAM8013 is available in the U-FLGA151-9 package.



## Features

- Ultra-Low EMI, -20dB Better Than FCC Class-B @ 300MHz
- High Efficiency up to 90% @1W with an 8Ω Speaker
- Shutdown Current <1µA
- 3W@10% THD Output with a 4Ω Load at 5V Supply
- Demanding Few External Components
- Superior Low Noise without Input
- Supply Voltage from 2.8V to 5.5 V
- Short-Circuit Protection
- Thermal Shutdown
- Available in the Space-Saving U-FLGA1515-9 Package
- Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)
- Halogen- and Antimony-Free. "Green" Device (Note 3)
- For automotive applications requiring specific change control (i.e. parts qualified to AEC-Q100/101/104/200, PPAP capable, and manufactured in IATF 16949 certified facilities), please <u>contact us</u> or your local Diodes representative. <u>https://www.diodes.com/quality/product-definitions/</u>

## Applications

- Cellular phones/smartphones
- MP4/MP3 devices
- GPS devices
- Digital photo frames
- Electronic dictionaries
- Portable game machines

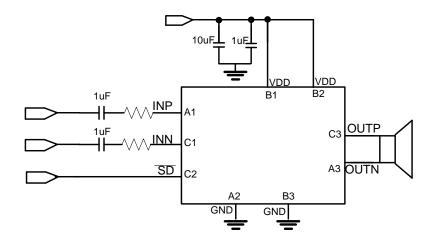
1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.

- 2. See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
- 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.
- 4. For packaging details, go to our website at https://www.diodes.com/design/support/packaging/diodes-packaging/.

Notes:



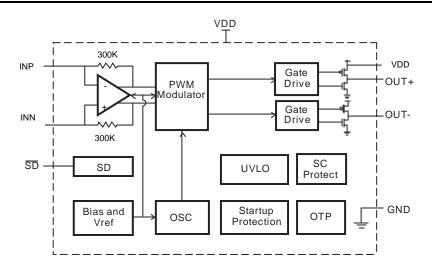
# **Typical Applications Circuit**



# **Pin Descriptions**

Pin Number	U-FLGA1515-9	Function	
A1	INP	Positive differential input	
A2	GND	Ground	
A3	OUTN	Negative BTL output	
B1	VDD	Power supply	
B2	VDD	Power supply	
B3	GND	Ground	
C1	INN	Negative differential input	
C2	SD	SD terminal to disable the chip	
C3	OUTP	Positive BTL output	

# **Functional Block Diagram**





# Absolute Maximum Ratings (@ T<sub>A</sub> = +25°C, unless otherwise specified.)

Parameter	Rating	Unit	
Supply Voltage (V <sub>DD</sub> )	6.0	V	
Input Voltage(IN+, IN-, CTRL)	-0.3 to V <sub>DD</sub> +0.3	V	
Storage Temperature	-65 to +150	°C	
Maximum Junction Temperature	150	°C	

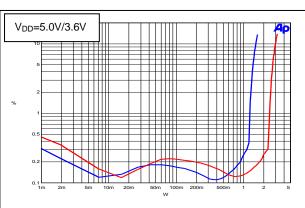
# **Recommended Operating Conditions** (@ T<sub>A</sub> = +25°C, unless otherwise specified.)

Symbol	Parameter	Min	Мах	Unit	
V <sub>DD</sub>	Supply Voltage	2.8	5.5	V	
T <sub>A</sub>	perating Ambient Temperature Range -25 +85		ŝ		
TJ	Junction Temperature Range	-40	+125		

## Electrical Characteristics (@ T<sub>A</sub> = +25°C, V<sub>DD</sub> = 5V, Gain = 18dB, R<sub>L</sub> = L (33µH)+R+L(33µH), unless otherwise noted.)

Symbol	Parameter	Test Conditions		Min	Тур	Max	Unit
V <sub>DD</sub>	Supply Voltage			2.8	_	5.5	V
		$V_{DD} = 5.0V$	_	3.0	_		
		THD+N = 10%, f = 1kHz, R = 4Ω	V <sub>DD</sub> = 3.6V	_	1.5	_	W
			$V_{DD} = 3.2V$	_	1.2	_	
		THD+N = 1%, f = 1kHz, R = 4Ω	$V_{DD} = 5.0V$	_	2.4	_	w
			V <sub>DD</sub> = 3.6V		1.25	_	
-			$V_{DD} = 3.2V$	_	1.0	_	
Po	Output Power		$V_{DD} = 5.0V$	_	1.75	_	
		THD+N = 10%, f = 1kHz, R = 8Ω	$V_{DD} = 3.6V$	_	0.90	_	W
			$V_{DD} = 3.2V$	_	0.70	_	
			$V_{DD} = 5.0V$	—	1.40	_	w
		THD+N = 1%, f = 1kHz, R = 8Ω	V <sub>DD</sub> = 3.6V	_	0.72	_	
			$V_{DD} = 3.2V$		0.60	_	
		$V_{DD} = 5.0V, Po = 1W, R = 8\Omega$		—	0.17	_	
		$V_{DD} = 3.6V, Po = 0.1W, R = 8\Omega$	f = 1kHz	_	0.16	_	%
	Total Harmonic	$V_{DD} = 3.2V, Po = 0.1W, R = 8\Omega$		_	0.14	_	
THD+N	Distortion Plus Noise	$V_{DD} = 5.0V, Po = 0.5W, R = 4\Omega$	f = 1kHz 0.16	_	0.14	_	%
	NOISE	$V_{DD} = 3.6V, Po = 0.2W, R = 4\Omega$		_	0.16	_	
		$V_{DD} = 3.2V, Po = 0.1W, R = 4\Omega$		0.17	_		
	Dewer Currely Dinele	ower Supply Ripple $V_{DD} = 3.6V$ , Inputs ac-grounded	f = 217Hz	—	-68	_	dB
PSRR	Rejection		f = 1kHz	—	-70	—	
	Појсскоп		f = 10kHz	—	-67	_	
Dyn	Dynamic Range	$V_{DD} = 5V$ , THD = 1%, R = 8 $\Omega$	f = 1kHz	—	95	_	dB
Vn	Output Noise	Inputs ac-grounded	No A weighting	—	170	_	μV
		1 5	A-weighting	—	130	_	
η	Efficiency $R_L = 8\Omega$ , THD = 10%	f = 1kHz		93	_	%	
1	,	$R_L = 4\Omega$ , THD = 10%		—	86	_	
lq	Quiescent Current	$V_{DD} = 5V$	No Load	—	5	_	mA
I <sub>SD</sub>	Shutdown Current	$V_{DD} = 2.8V$ to 5V	CTRL = 0V	—		1	μA
R <sub>DS(ON)</sub>	Static Drain-to Source	High Side PMOS, I = 500mA	$V_{DD} = 5.0V$	—	325	_	mΩ
· · ·		Low Side NMOS, I = 500mA	$V_{DD} = 5.0V$	—	200	_	mΩ
fsw	Switching Frequency	$V_{DD} = 2.8V$ to 5V		—	400	_	kHz
Gv	Closed-loop Gain	$V_{DD} = 2.8V$ to 5V		—	300K/R <sub>IN</sub>	_	V/V
RIN	Input Impedance	$V_{DD} = 2.8V$ to 5V		—	28.5	—	ΚΩ
Vo	Output Offset Voltage	Input ac-ground, V <sub>DD</sub> = 5V		—	—	20	mV
VIH	SD Input High Voltage	$V_{DD} = 5V$		1.4	—	—	v
VIL	SD Input Low Voltage	$V_{DD} = 5V$		_	—	1.0	v

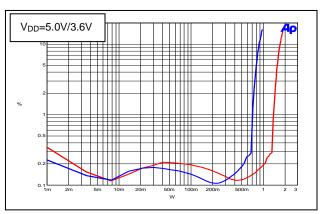




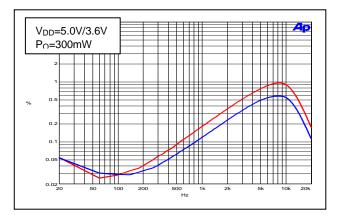
### THD+N Vs. Output Power (RL=4 $\Omega$ )

### THD+N Vs. Output Power (RL=8 $\Omega$ )

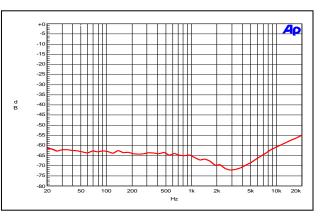
PAM8013



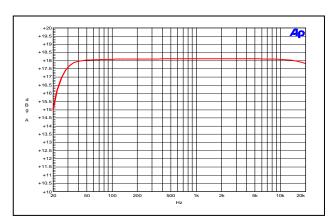
THD+N Vs. Frequency



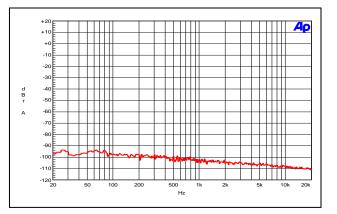




Frequency Response



Noise Floor





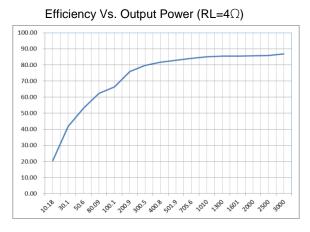
# **Performance Characteristics** (@ $T_A = +25^{\circ}C$ , $V_{DD} = 5V$ , Gain = 18dB, $R_L = L$ (33µH)+R+L(33µH), unless otherwise noted.)

30.00

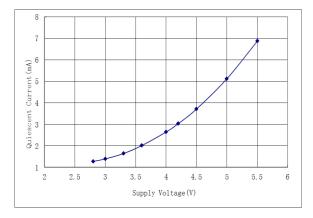
20.00

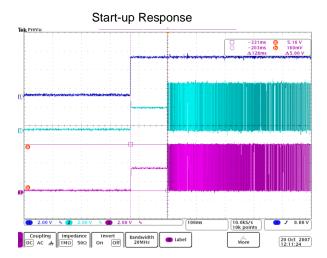
10.00

0.00



Quiescent Current vs. Supply Voltage

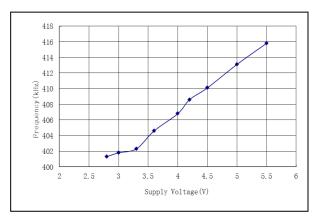


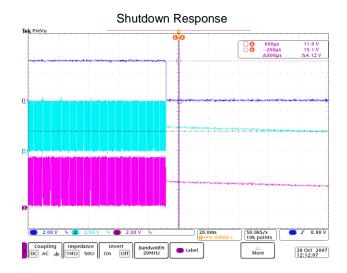


Efficiency Vs. Output Power (RL=8Ω)

OSC Frequency vs. Supply Voltage

1013 301 10h 100 189 102 30 101 605 803 100 130 150 100 10h







## **Application Information**

### Input Capacitors (Ci)

In the typical application, an input capacitor, Ci, is required to allow the amplifier to bias the input signal to the proper DC level for optimum operation. In this case, Ci and the minimum input impedance Ri form is a high-pass filter with the corner frequency determined in the following equation:

$$fc = \frac{1}{(2\pi RiCi)}$$

It is important to consider the value of Ci as it directly affects the low frequency performance of the circuit; for example, when Ri is  $150k\Omega$  and the specification calls for a flat bass response down to 150Hz. The equation is reconfigured as follows:

$$\mathrm{Ci} = \frac{1}{\left(2\pi\mathrm{Rifc}\right)}$$

When input resistance variation is considered, the Ci is 7nF and one would likely choose a value of 10nF. A further consideration for this capacitor is the leakage path from the input source through the input network (Ci, Ri + Rf) to the load. This leakage current creates a DC offset voltage at the input to the amplifier that reduces useful headroom, especially in high gain applications. For this reason, a low-leakage tantalum or ceramic capacitor is the best choice. When polarized capacitors are used, the positive side of the capacitor should face the amplifier input in most applications as the DC level is held at VDD/2, which is likely higher than the source DC level. Please note that it is important to confirm the capacitor polarity in the application.

### **Decoupling Capacitor (CS)**

The PAM8013 is a high-performance CMOS audio amplifier that requires adequate power supply decoupling to ensure that the output total harmonic distortion (THD) is as low as possible. Power supply decoupling also prevents the oscillations causing by long lead length between the amplifier and the speaker.

The optimum decoupling is achieved by using two different types of capacitors that target on different types of noise on the power supply leads. For the best operation with higher frequency transients, spikes, or digital hashes on the line, a good low equivalent-series-resistance (ESR) ceramic capacitor, typically  $1\mu$ F, is placed as close as possible to the device's V<sub>DD</sub> pin. For filtering lower frequency noise signals, a large ceramic capacitor of  $10\mu$ F or greater is recommended to be placed near the audio power amplifier.

### How to Reduce EMI

Most applications require a ferrite bead filter for EMI elimination, as shown in Figure 1. The ferrite filter reduces EMI around 1MHz and higher. When selecting a ferrite bead, choose one with high impedance at high frequencies, but low impedance at low frequencies.

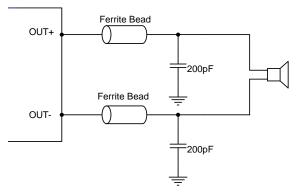


Figure 1: Ferrite Bead Filter to Reduce EMI



### Application Information (continued)

### **Shutdown Operation**

In order to reduce power consumption while not in use, the PAM8013 contains shutdown-circuitry amplifier-off when logic-low is placed on the CTRL pin. By switching the shutdown pin connected to GND, the PAM8013 supply current draw will be minimized in idle mode.

### Undervoltage Lock-out (UVLO)

The PAM8013 incorporates circuitry designed to detect low supply voltage. When the supply voltage drops to 2.0V or below, the PAM8013 goes into a state of shutdown. Only when VDD higher than 2.2V will the device come out of its shutdown state and restore to normal function.

### Short-Circuit Protection (SCP)

The PAM8013 has short-circuit protection circuitry on the outputs to prevent the device from damage when output-to-output shorts or output-to-GND shorts occur. When a short circuit occurs, the device immediately goes into shutdown state. Once the short is removed, the device will be reactivated.

#### **Overtemperature Protection (OTP)**

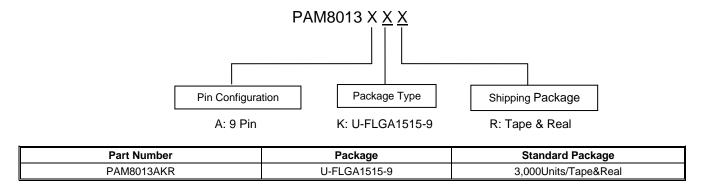
Thermal protection on the PAM8013 prevents the device from damage when the internal die temperature exceeds 150°C. There is a 15°C tolerance on this trip point from device to device. Once the die temperature exceeds the set point, the device will enter the shutdown state and the outputs are disabled. This is not a latched fault. The thermal fault is cleared once the temperature of the die decreases by 40°C. This large hysteresis will prevent motor-boating sound well and the device will begin normal operation at this point with no external system interaction.

### POP and Click Circuitry

The PAM8013 contains circuitry to minimize turn-on and turn-off transients, or "click and pops", where turn-on refers to either power supply turnon or device recovery from shutdown mode. When the device is turned on, the amplifiers are internally muted. An internal current source ramps up the internal reference voltage. The device will remain in mute mode until the reference voltage reaches half-supply voltage, 1/2 V<sub>DD</sub>. As soon as the reference voltage is stable, the device will begin full operation. For the best power-off pop performance, the amplifier should be set in shutdown mode prior to removing the power supply voltage.

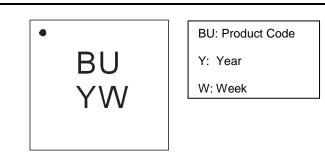


## **Ordering Information**



# **Marking Information**

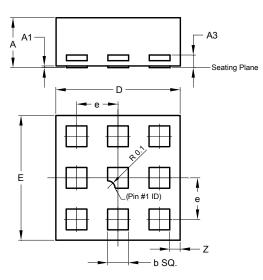
U-FLGA1515-9





# Package Outline Dimensions (All dimensions in mm.)

Please see http://www.diodes.com/package-outlines.html for the latest version.



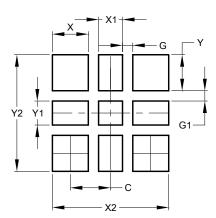
	U-FLGA1515-9				
Dim	Min	Max	Тур		
Α	0.55	0.65	0.60		
A1	0	0.05	0.02		
A3	0.13 BSC				
b	0.20	0.30	0.25		
D	1.45	1.55	1.50		
Е	1.45	1.55	1.50		
е	0.50 BSC				
Z	0.125 BSC				
All D	All Dimensions in mm				

# **Suggested Pad Layout**

Please see http://www.diodes.com/package-outlines.html for the latest version.

### U-FLGA1515-9

U-FLGA1515-9



Dimensions	Value (in mm)	
С	0.587	
G	0.150	
G1	0.150	
Х	0.525	
X1	0.350	
X2	1.700	
Ŷ	0.525	
Y1	0.350	
Y2	1.700	

## **Mechanical Data**

- Moisture Sensitivity: Level 1 per J-STD-020
- Terminals: Finish—NiPdAu, Solderable per J-STD-002, Test B1 (4)
- Weight: 2.4mg (Approximate)



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