















**PRIMARY SIDE CONTROL IC FOR OFF-LINE BATTERY CHARGERS AP3708N**

**Typical Performance Characteristics (Continued)**

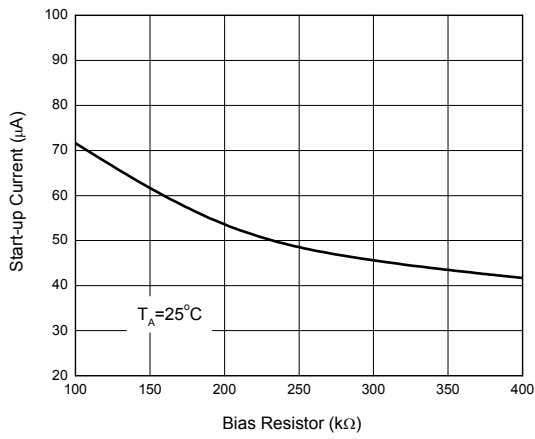


Figure 8. Start-up Current vs. Bias Resistor

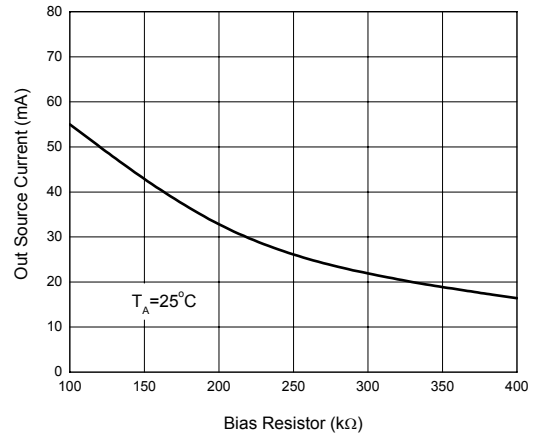


Figure 9. OUT Source Current vs. Bias Resistor

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**Operation Description**

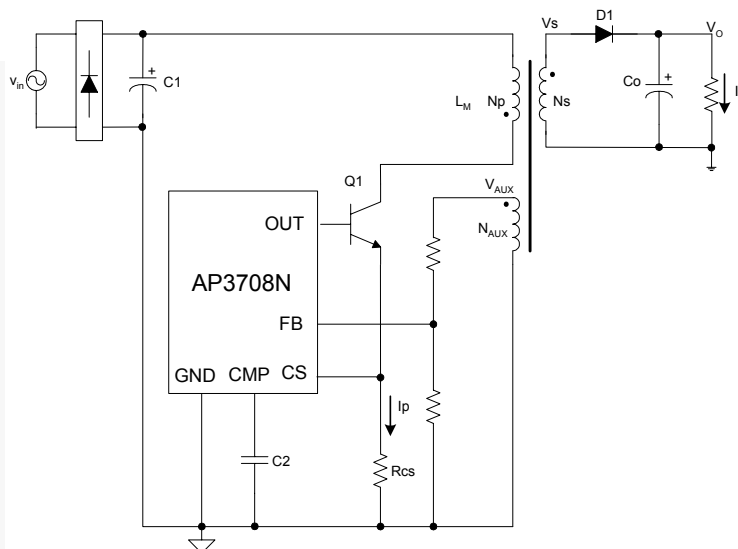


Figure 10. Simplified Flyback Converter Controlled by AP3708N

Figure 10 illustrates a simplified flyback converter controlled by AP3708N.

The energy stored in the magnetizing inductance  $L_M$  each cycle is therefore:

**Constant Primary Peak Current**

The primary current  $i_p(t)$  is sensed by a current sense resistor  $R_{CS}$  as shown in Figure 10.

$$E_g = \frac{1}{2} \times L_M \times I_{pk}^2 \quad \dots\dots(3)$$

The current rises up linearly at a rate of:

So the power transferring from the input to the output is given by:

$$\frac{di_p(t)}{dt} = \frac{v_g(t)}{L_M} \quad \dots\dots(1)$$

$$P = \frac{1}{2} \times L_M \times I_{pk}^2 \times f_{SW} \quad \dots\dots(4)$$

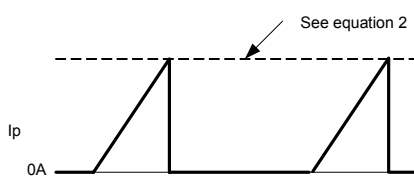


Figure 11. Primary Current Waveform

As illustrated in Figure 11, when the current  $i_p(t)$  rises up to  $I_{pk}$ , the switch Q1 turns off. The constant peak current is given by:

$$I_{pk} = \frac{V_{cs}}{R_{cs}} \quad \dots\dots(2)$$

where  $f_{SW}$  is the switching frequency. When the peak current  $I_{pk}$  is constant, the output power depends on the switching frequency  $f_{SW}$ .

**Constant Voltage Operation**

The AP3708N captures the auxiliary winding feedback voltage at FB pin and operates in constant-voltage (CV) mode to regulate the output voltage. Assuming the secondary winding is master, the auxiliary winding is slave during the D1 on-time. The auxiliary voltage is given by:

$$V_{AUX} = \frac{N_{AUX}}{N_S} \times (V_o + V_d) \quad \dots\dots(5)$$



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**Operation Description (Continued)**

where  $V_d$  is the diode forward drop voltage.

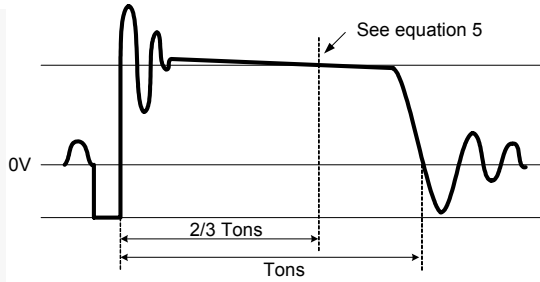


Figure 12. Auxiliary Voltage Waveform

The output voltage is different from the secondary voltage in a diode forward drop voltage that depends on the current. If the secondary voltage is always detected at a fixed secondary current, the difference between the output voltage and the secondary voltage will be a fixed  $V_d$ . The voltage detection point is at two-thirds of the D1 on-time. The CV loop control function of AP3708N then generates a D1 off-time to regulate the output voltage.

**Constant Current Operation**

Figure 13 shows the secondary current waveforms.

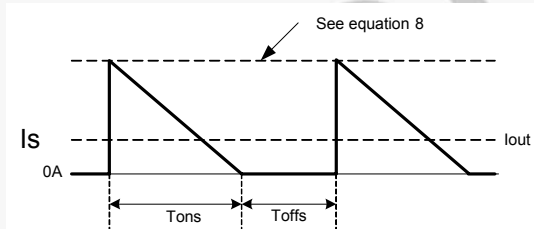


Figure 13. Secondary Current Waveform

In CC operation, the CC loop control function of AP3708N will keep a fixed proportion between D1 on-time  $T_{ons}$  and D1 off-time  $T_{offs}$  by discharging or charging the capacitance connected in CMP pin. The fixed proportion is

$$\frac{T_{ons}}{T_{offs}} = \frac{4}{3} \quad \dots\dots(6)$$

The relationship between the output constant-current and secondary peak current  $I_{pks}$  is given by:

$$I_{out} = \frac{1}{2} \times I_{pks} \times \frac{T_{ons}}{T_{ons} + T_{offs}} \quad \dots\dots(7)$$

At the instant of D1 turn-on, the primary current transfers to the secondary at an amplitude of:

$$I_{pks} = \frac{N_p}{N_s} \times I_{pk} \quad \dots\dots(8)$$

Thus the output constant-current is given by:

$$I_{out} = \frac{1}{2} \times \frac{N_p}{N_s} \times I_{pk} \times \frac{T_{ons}}{T_{ons} + T_{offs}} = \frac{2}{7} \times \frac{N_p}{N_s} \times I_{pk} \quad \dots\dots(9)$$

**Leading Edge Blanking**

When the power switch is turned on, a turn-on spike will occur on the sense-resistor. To avoid false-termination of the switching pulse, a 430ns leading-edge blanking is built in. During this blanking period, the current sense comparator is disabled and the gate driver can not be switched off.

**CCM Protection**

The AP3708N is designed to operate in discontinuous conduction mode (DCM) in both CV and CC modes. To avoid operating in continuous conduction mode (CCM), the AP3708N detects the falling edge of the FB input voltage on each cycle. If a 0.1V falling edge of FB is not detected, the AP3708N will stop switching.

**OVP & OckP**

The AP3708N includes output over-voltage protection (OVP) and open circuit protection (OckP) circuitry as shown in Figure 14. If the voltage at FB pin exceeds 8V, 100% above the normal detection voltage, or the -0.7V falling edge of the FB input can not be monitored, the AP3708N will immediately shut off and enter hiccup mode. The AP3708N sends out a fault detection pulse every 32ms in hiccup mode until the fault has been removed.

**PRIMARY SIDE CONTROL IC FOR OFF-LINE BATTERY CHARGERS AP3708N**

**Operation Description (Continued)**

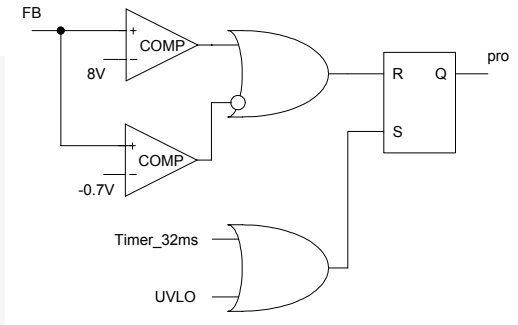


Figure 14. OVP and OCKP Function Block

Through the internal RA and RB, the FB voltage can be compensated by the Vout1, the compensation voltage is 0.4V when full load switch frequency is 60kHz.

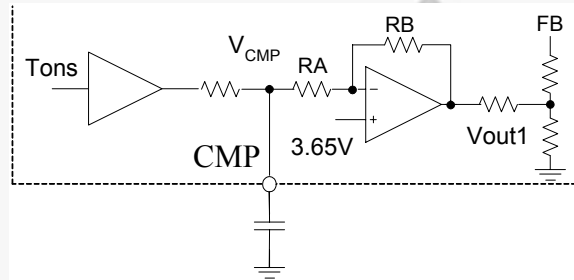


Figure 15. Output Cable Compensation Function Block

**Output Cable Compensation**

The AP3708N integrates the output cable compensation circuitry as shown in Figure 15. Tons shows the variation for FB flyback voltage. Tons can be converted to a DC voltage by a low-pass filter. When system load current Iout changed from open load to full load Iload, The amplified voltage Vout1 through a rail-to-rail operation amplifier is obtained:

$$V_{OUT1} = \left(1 + \frac{RB}{RA}\right) \times 3.65V - \frac{RB}{RA} \times V_{CMP} \quad \dots\dots(10)$$

OBSOLETE - PART DISCONTINUED



PART OBSOLETE

Preliminary Datasheet

**PRIMARY SIDE CONTROL IC FOR OFF-LINE BATTERY CHARGERS AP3708N**

**Typical Application**

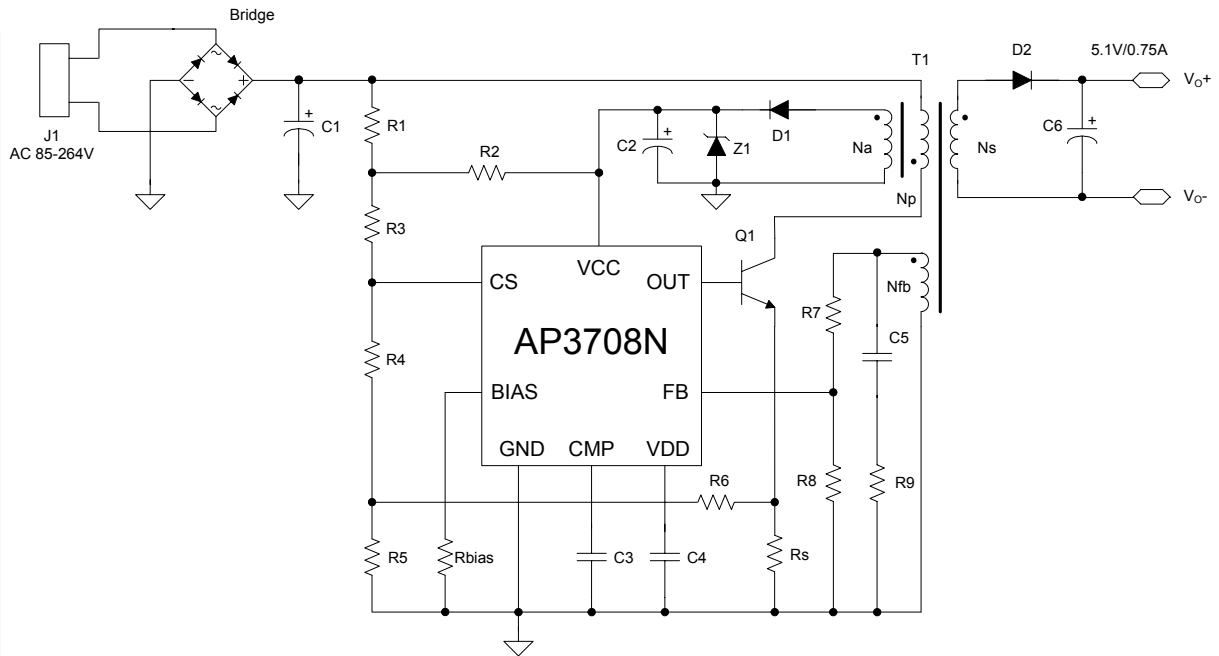


Figure 16. 5V/1A Output for Battery Charger of Mobile Phone



PART OBSOLETE

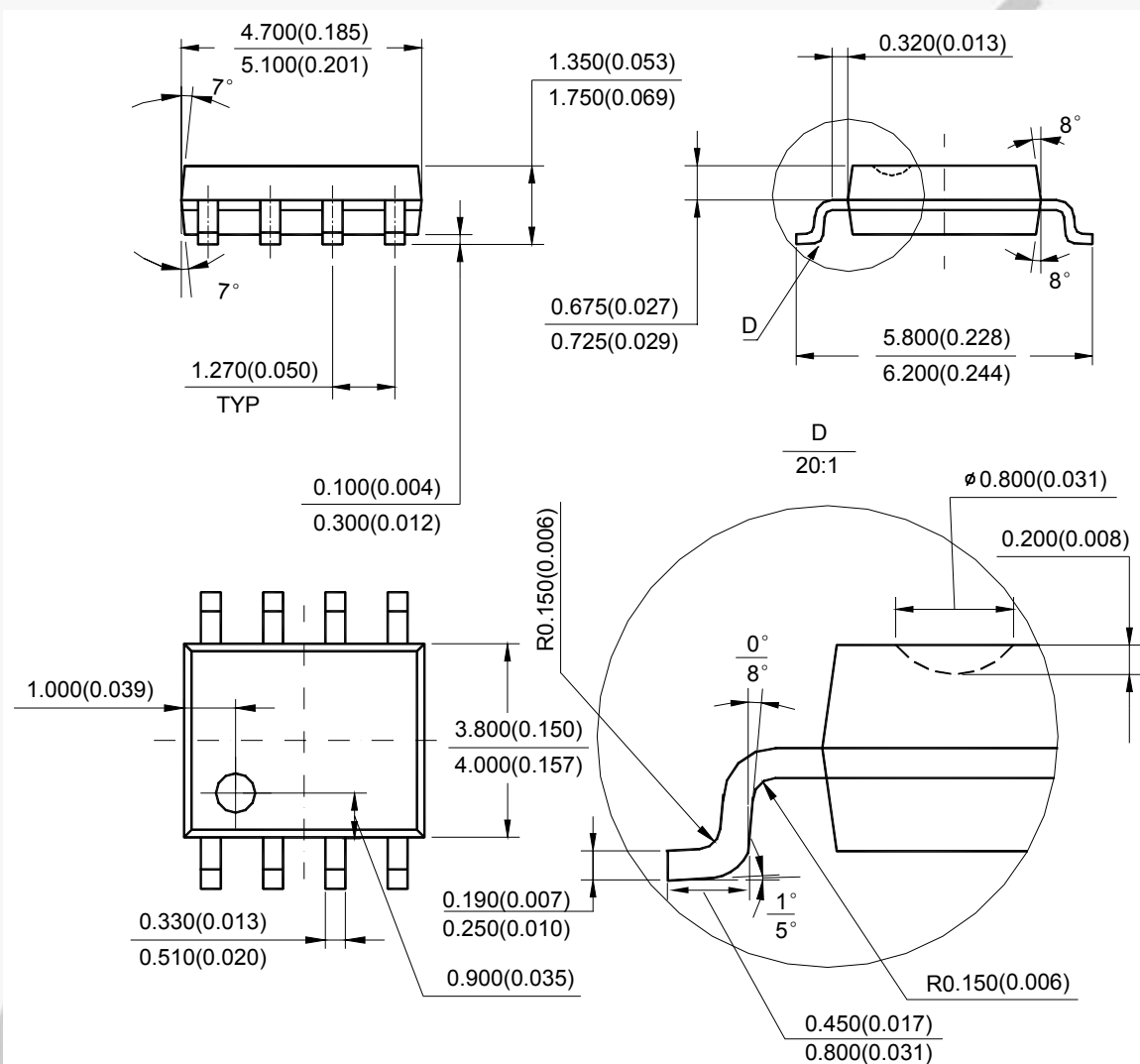
Preliminary Datasheet

**PRIMARY SIDE CONTROL IC FOR OFF-LINE BATTERY CHARGERS AP3708N**

**Mechanical Dimensions**

SOIC-8

Unit: mm(inch)



Note: Eject hole, oriented hole and mold mark is optional.

PART OBSOLETE



**BCD Semiconductor Manufacturing Limited**

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