

# AN1161 BCR430UW6 Application Information

This document details the application and use of the Diodes BCR430UW6 low dropout linear LED driver. It should be used in conjunction with the BCR430UW6 data sheet which details the full electrical specification of the BCR430UW6.

Sections of this document include external resistor selection, power dissipation and over temperature protection (OTP), PWM dimming, SPICE modeling and the BCR430UW6 evaluation board which allows engineers to appraise and evaluate the BCR430UW6 when designing reliable and robust LED driver solutions for LED strings.

Applications for the BCR430UW6 include:

- Architectural lighting
- Retail lighting
- Automotive lighting
- Strip lighting



Figure 1: Typical Application



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#### 1 Introduction

The BCR430U is a monolithically integrated linear LED controller designed to function as a Constant Current Regulator (CCR) for linear LED driving. The device operates over a voltage range from 5V to 42V and regulates the output LED current up to 100mA, set by an external resistor. It is designed for driving LEDs in strings and will reduce current at increasing temperatures to self-protect. The low voltage drop during current regulation allows efficient driving of LED strings with a range of forward voltages and supply voltage tolerances Inbuilt Over Temperature Protection (OTP) gradually reduces the output current when the chip temperature reaches 125 °C.



# **2 Rext Calculation**

The LED current *lout* is set by a resistor *Rext* connected from REXT to GND. The REXT pin is driven by a 900mV reference and the resultant REXT current is mirrored to the Driver and OTP stages.



Figure 2: Rext Calculation

There is an exponential relationship between *Rext* and *lout* expressed by the formula:

 $Rext = \left(\frac{Iout}{610.53}\right)^{-1.005632}$ 

where: Rext and lout are expressed in  $k\Omega$  and mA



Figure 3: Rext vs lout



#### **3 Power Dissipation**

Power dissipation occurs in several areas and varies according to the operating conditions, however the major influence on power dissipation is the  $V_{DS}$  voltage of the output stage. An example of power dissipation and Junction temperature rise is shown below:

Typical Power Dissipation when  $V_{DS} = 4V$ ,  $Rext = 13k\Omega$ 

Typical Power Dissipation $V_{DS} = 4V$ , $Rext = 13k\Omega$ refer to <i>figure 2</i>			
Parameter	Current	Power	
Quiescent Current	67µA	2.8mW	
Reference Current	69µA	2.9mW	
Regulation current	24µA	1mW	
Operating Current	125µA	5.3mW	
Output Current	50mA	211mW	

Figure 4: Power Dissipation

The **BCR430UW6.xlxs** calculator tool can be used to easily set *Rext* value and show power consumption under various operating conditions.



#### 4 Over Temperature Protection (OTP)

Over temperature protection is included to safeguard the device under fault conditions such as short circuit LED's or high ambient temperatures. As the chip temperature increases above 125 °C the output is reduced at a rate of *Isetpoint* / 30 mA per °C rise. The smooth operation maximises LED current and removes flickering.



As  $V_{DS}$  increases in the output stage, so will power dissipation and chip temperature. The effects of OTP can be seen in the following simulation which also shows how the ambient temperature and thermal resistance *RthJA* can be adjusted in the SPICE model.



Figure 6: Gradual reduction of output current as Rds losses increase



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#### **PWM Dimming** 5

Dimming is most efficiently carried out using PWM techniques. This can be achieved by switching either the REXT pin or VS pin.

Switching the VS supply reduces the connection requirements in multiple device systems such as large signs but requires attention to miller effect and transients during power connection.

Note that human visual light level perception and true optical power have a logarithmic relationship where the perceived level is based on the output power ratio:

Perceived power =  $\sqrt{(Output power / Maximum power)}$ 

and

Perceived power (%) = 100  $\times \sqrt{(\text{Output power (%)} / 100)}$ 

Converting this to a PWM dimming level:

Perceived power =  $\sqrt{(Maximum power * Duty Cycle)}$ 

Transposing to solve for Output power ratio

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Output power ratio = Perceived power<sup>2</sup>
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#### and for PWM

Duty Cycle = Perceived power<sup>2</sup> / Maximum power



Figure 7: Perceived power vs Actual power



#### 5.1 PWM Safe operating area (SOA)

There are various safety standards for PWM dimming to prevent physiological and safety issues such as IEEE1789. The safe operating area graph from IEEE1789 is reproduced below.



Figure 8: IEEE1789 safe operating area

#### 5.2 PWM Frequency and pulse width

The maximum frequency of operation depends on the dimming requirements and the switching time. The Datasheet specifies worst case tr =  $14\mu$ s, tf =  $3\mu$ s so we may consider a minimum duty cycle  $40\mu$ s as reasonable.

If we want to dim to a perceived level of 10% full brightness the actual level required is:

Actual Power = (Perceived Power)<sup>2</sup> \* Maximum Power

 $= 0.1^2 * Maximum Power$ 

= 0.01 \* Maximum Power

If the minimum duty cycle is 40µs then:

100% duty cycle time = 40µs / 0.01 = 4ms = 250Hz

The table below shows the exponential nature of perceived brightness dimming

Required Cycle Time to achieve Required Dimming at 40uS Duty cycle					
Perceived %	Percieved Ratio	Output Ratio	Output %	Cycle Time(us)	Freq (Hz)
100	1	1	100	40.000	25000
90	0.9	0.81	81	49.383	20250
80	0.8	0.64	64	62.500	16000
70	0.7	0.49	49	81.633	12250

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60	0.6	0.36	36	111.111	9000
50	0.5	0.25	25	160.000	6250
40	0.4	0.16	16	250.000	4000
30	0.3	0.09	9	444.444	2250
20	0.2	0.04	4	1000.000	1000
10	0.1	0.01	1	4000.000	250
9	0.09	0.0081	0.81	4938.272	202.5
8	0.08	0.0064	0.64	6250.000	160
7	0.07	0.0049	0.49	8163.265	122.5
6	0.06	0.0036	0.36	11111.111	90
5	0.05	0.0025	0.25	16000.000	62.5
4	0.04	0.0016	0.16	25000.000	40
3	0.03	0.0009	0.09	44444.444	22.5
2	0.02	0.0004	0.04	100000.000	10
1	0.01	0.0001	0.01	400000.000	2.5

Figure 9: min frequencies and SOA

# 5.2 REXT Dimming

The most efficient dimming is by PWM of the REXT pin. The value of Rext is calculated for maximum current requirement at 100% duty cycle. Using a GPIO to pull Rext low will turn on the output. The output impedance of the GPIO will have some impact on the required value of Rext. When the GPIO is high (above 900mV) the output will turn off. The REXT pin will clamp at 5V.



Figure 10: REXT Dimming by GPIO

### 5.3 Supply Voltage Dimming



PWM dimming can be achieved by modulating the power supply. If the power supply has a fast rise time the drain-gate capacitance of the output transistor may cause the transistor to turn on for  $\sim$  650ns before the control circuit has stabilized. This can be mitigated by increasing the rise time of the power supply to greater than 5µs.



Figure 11: Power Supply PWM Dimming



#### 6 ESD Protection

ESD protection is 2kV HBM. This can be extended to 6kV HBM by placing a  $1k\Omega$  resistor in series with the VCC pin.



Figure 12: 6kV ESD Protection

# 7 SPICE Model

A SPICE model BCR430UW6.spice.txt is available on the Diodes website. This is a behavioural model that includes OTP modelling.

The model has extra inputs to set ambient temperature and junction-ambient thermal resistance (RthJA).



	Model Pin Descriptions			
PIN	NAME	FUNCTION		
1	RthJA	Input sets thermal resistance. 1V = 1 °C/W		
2	GND	0V		
3	REXT	Current setpoint resistor connection		
4	VS	V+ supply 0 - 42V		
5	Tamb	Input sets ambient temperature. 1 V = 1 °C		
6	OUT	Output current sink		

Figure 13: Spice Model



Figure 13: Example of OTP operation as  $l^2 R_{DS}$  losses increase with Vout



### 8 Calculator

A calculator **BCR430UW6.Calculator.xlsx** is available on the Diodes website. The calculator enables quick calculation of Rext, power dissipation, operating temperature and PWM ranges.



Figure 14: BCR430UW6 Calculator



#### 9 Evaluation Board

An evaluation board is available that allows evaluation of the BCR430UW6 with an onboard string of 8 Liteon LEDs or external LEDs of your choice. Constant current or PWM is selectable. A testpoint is included to facilitate Vout measurement to verify output losses ad LED performance.

#### 9.1 Technical specification

Input Voltage	5 – 42 Vdc
Default Rext	12 kΩ
Default LED current	50 mA
Output Current Range	5 -100 mA
Dimensions	35 mm x 55 mm

#### 9.2 Circuit



Figure 15: Circuit of BCR430UW6 Evaluation Board

P1 jumper is used to select on-board LED's or external LED's.

P2 jumper is used to select Rext resistor connected to GND (50mA) or to PWM pin for external dimming 5V dc. max

ZD1 protects against over voltage or negative PWM signal

TP1 accepts oscilloscope spring probe to measure OUT voltage

R1 also has PTH pads and can be removed or replaced with a low value for current monitoring or ESD testing



#### 9.3 PCB

The PCB is 2 layer 1.5mm FR4 1oz copper. Thermal relief is provided for LED's and BCR430UW6.

The LED's and BCR430UW6 are connected to copper areas top and bottom of 200 mm<sup>2</sup> per device for heat dissipation.



Figure 16: Top layer

Figure 17: Bottom layer

Figure 18: 3D

#### 9.4 Bill of Materials

Bill of Materials		
Designator	Quantity	Value
C1	1	10n 0805 50V
D1-8	8	LTW-5630AQL30
P1, P2	2	3 pin header
R1	1	0R 1206 200V
Rext	1	12k 0805 1% 150V
U1	1	BCR430UW6
ZD1	N.F.	BZX84C4V7-7-F



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